DUAL UART WITH 64-BYTE FIFO

Check for Samples: TL16C752C

FEATURES

- ST16C654/654D Pin Compatible With Additional Enhancements (PFB Package Only)
- Supports up to 24-MHz Crystal Input Clock (1.5 Mbps)
- Supports up to 48-MHz Oscillator Input Clock (3 Mbps) for 5-V Operation
- Supports up to 32-MHz Oscillator Input Clock (2 Mbps) for 3.3-V Operation
- Supports up to 24-MHz Input Clock (1.5 Mbps) for 2.5-V Operation
- Supports up to 16-MHz Input Clock (1 Mbps) for 1.8-V Operation
- 64-Byte Transmit FIFO
- 64-Byte Receive FIFO With Error Flags
- Programmable and Selectable Transmit and Receive FIFO Trigger Levels for DMA and Interrupt Generation
- Programmable Receive FIFO Trigger Levels for Software/Hardware Flow Control
- Software/Hardware Flow Control
 - Programmable Xon/Xoff Characters
 - Programmable Auto-RTS and Auto-CTS
- Optional Data Flow Resume by Xon Any Character
- DMA Signaling Capability for Both Received and Transmitted Data on PN Package

- RS-485 Mode Support
- Support 1.8-V, 2.5-V, 3.3-V, or 5-V Supply
- Characterized for Operation From –40°C to 85°C, Available in Commercial and Industrial Temperature Grades
- Software-Selectable Baud-Rate Generator
- Prescalable Provides Additional Divide-by-4 Function
- Programmable Sleep Mode
- Programmable Serial Interface Characteristics
 - 5-, 6-, 7-, or 8-Bit Characters
 - Even, Odd, or No Parity Bit Generation and Detection
 - 1-, 1.5-, or 2-Stop Bit Generation
- False Start Bit Detection
- Complete Status Reporting Capabilities in Both Normal and Sleep Mode
- Line Break Generation and Detection
- Internal Test and Loopback Capabilities
- Fully Prioritized Interrupt System Controls
- Modem Control Functions (CTS, RTS, DSR, DTR, RI, and CD)
- IrDA Capability

DESCRIPTION

The TL16C752C is a dual universal asynchronous receiver/transmitter (UART) with 64-byte FIFOs, automatic hardware/software flow control, and data rates up to 3 Mbps. It incorporates the functionality of four UARTs, each UART having its own register set and FIFOs. The four UARTs share only the data bus interface and clock source, otherwise they operate independently. Another name for the UART function is asynchronous communications element (ACE), and these terms are used interchangeably. The bulk of this document describes the behavior of each ACE, with the understanding that four such devices are incorporated into the TL16C752C. The TL16C752C offers enhanced features. It has a transmission control register (TCR) that stores received FIFO threshold level to start/stop transmission during hardware and software flow control. With the FIFO RDY register, the software gets the status of TXRDY/RXRDY for all four ports in one access. On-chip status registers provide the user with error indications, operational status, and modem interface control. System interrupts may be tailored to meet user requirements. An internal loopback capability allows onboard diagnostics.

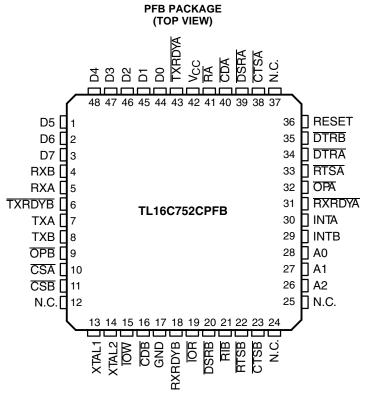


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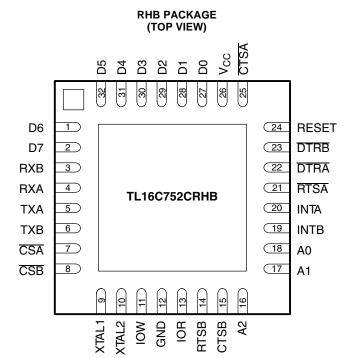
Each UART transmits data sent to it from the peripheral 8-bit bus on the TX signal and receives characters on the RX signal. Characters can be programmed to be 5, 6, 7, or 8 bits. The UART has a 64-byte receive FIFO and transmit FIFO and can be programmed to interrupt at different trigger levels. The UART generates its own desired baud rate based upon a programmable divisor and its input clock. It can transmit even, odd, or no parity and 1-, 1.5-, or 2-stop bits. The receiver can detect break, idle or framing errors, FIFO overflow, and parity errors. The transmitter can detect FIFO underflow. The UART also contains a software interface for modem control operations, and software flow control and hardware flow control capabilities.

The TL16C752C is available in a 32-pin QFN (RHB) package. A 48-pin QFP (PFB) package will be available in late 2008.



N.C. - No internal connection





NOTE: The 32-pin RHB package does not provide access to $\overline{\text{DSRA}}$, $\overline{\text{DSRB}}$, $\overline{\text{RIA}}$, $\overline{\text{RIB}}$, $\overline{\text{CDA}}$, $\overline{\text{CDB}}$ inputs or $\overline{\text{OPA}}$, $\overline{\text{OPB}}$, $\overline{\text{RXRDYA}}$, $\overline{\text{RXRDYA}}$, $\overline{\text{RXRDYA}}$ outputs.

TERMINAL FUNCTIONS

TERMINAL					
NAME	N	Ю.	1/0	DESCRIPTION	
NAME	PFB	RHB			
A0	28	18	I	Address bit 0 select. Internal registers address selection. Refer to Table 9 for Register Address Map.	
A1	27	17	I	Address bit 1 select. Internal registers address selection. Refer to Table 9 for Register Address Map.	
A2	26	16	I	Address bit 2 select. Internal registers address selection. Refer to Table 9 for Register Address Map.	
CDA, CDB,	40, 16		I	Carrier detect (active low). These inputs are associated with individual UART channels A through B. A low on these pins indicates that a carrier has been detected by the modem for that channel.	
CSA, CSB,	10, 11	7, 8	I	Chip select A and B (active low). These pins enable data transfers between the user CPU and the TL16C752C for the channel(s) addressed. Individual UART sections (A, E, D) are addressed by providing a low on the respective CSA through CSD pin.	
CTSA, CTSB,	38, 23	25, 15	I	Clear to send (active low). These inputs are associated with individual UART channels A and B. A low on the CTS pins indicates the modem or data set is ready to accept transmit data from the TL16C752C. Status can be checked by reading MSR[4]. These pins only affect the transmit and receive operations when auto CTS function is enabled through the enhanced feature register (EFR[7]), for hardware flow control operation.	
D0-D4, D5-D7	44–48, 1–3	27–31 32, 1, 2	I/O	Data bus (bidirectional). These pins are the eight-bit, 3-state data bus for transferring information to or from the controlling CPU. D0 is the least significant bit and the first data bit in a transmit or receive serial data stream.	
DSRA, DSRB,	39, 20		I	Data set ready (active low). These inputs are associated with individual UART channels A through B. A low on these pins indicates the modem or data set is powered on and is ready for data exchange with the UART.	
DTRA, DTRB,	34, 35	22, 23	0	Data terminal ready (active low). These outputs are associated with individual UART channels A through B. A low on these pins indicates that the TL16C752C is powered on and ready. These pins can be controlled through the modem control register. Writing a 1 to MCR[0] sets the DTR output to low, enabling the modem. The output of these pins is high after writing a 0 to MCR[0], or after a reset. These pins can also be used in the RS-485 mode to control an external RS-485 driver or transceiver.	

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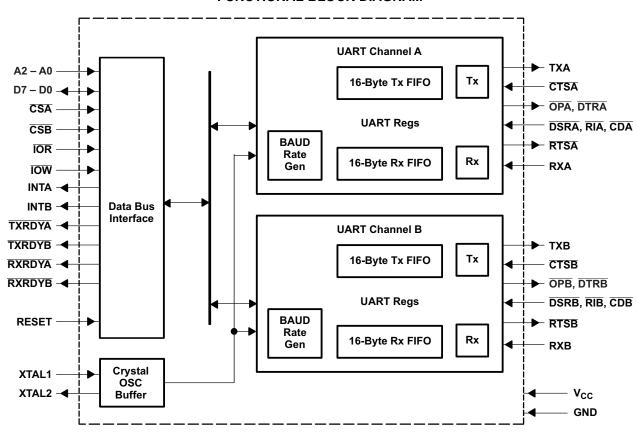


TERMINAL FUNCTIONS (continued)

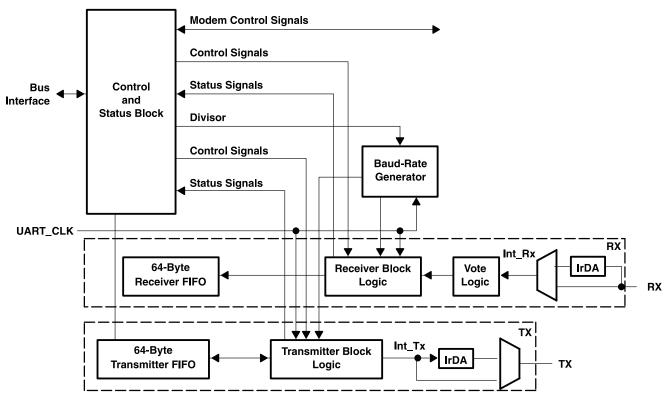
TERMINAL						
NO.		NO. I/O		DESCRIPTION		
NAME	PFB	RHB				
GND	17	12	Pwr	Power signal and power ground		
ĪNTĀ, ĪNTB,	30, 29	20, 19	0	Interrupt A and B (active high). These pins provide individual channel interrupts, INTA-D. INTA-D are enabled when MCR[3] is set to a 1, interrupts are enabled in the interrupt enable register (IER) and when an interrupt condition exists. Interrupt conditions include: receiver errors, available receiver buffer data, transmit buffer empty, or when a modem status flag is detected. INTA-D are in the high-impedance state after reset.		
ĪOR	19	13	I	Read input (active low strobe). A valid low level on IOR loads the contents of an internal register defined by address bits A0–A2 onto the TL16C752C data bus (D0–D7) for access by an external CPU.		
ĪŌW	15	11	I	Write input (active low strobe). A valid low level on $\overline{\text{IOW}}$ transfers the contents of the data bus (D0–D7) from the external CPU to an internal register that is defined by address bits A0–A2.		
NC	12, 24 35, 37			No internal connection		
OPA, OPB	32, 9		0	User defined outputs. This function is associated with individual channels A and B. The state of these pins is defined by the user through the software settings of the MCR register, bit 3. INTA-B are set to active mode and \overline{OP} to a logic 0 when the MCR-3 is set to a logic 1. INTA-B are set to the 3-state mode and \overline{OP} to a logic 1 when MCR-3 is set to a logic 0. See bit 3, modem control register (MCR bit 3). The output of these two pins is high after reset.		
RESET	36	24	I	Reset. RESET resets the internal registers and all the outputs. The UART transmitter output and the receiver input are disabled during reset time. See TL16C752C external reset conditions for initialization details. RESET is an active high input.		
RIA, RIB,	41, 21		I	Ring indicator (active low). These inputs are associated with individual UART channels A and B. A logic low on these pins indicates the modem has received a ringing signal from the telephone line. A low-to-high transition on these input pins generates a modem status interrupt, if enabled. The state of these inputs is reflected in the modem status register (MSR).		
RTSA, RTSB,	33, 22	21, 14	0	Request to send (active low). These outputs are associated with individual UART channels A through D. A low on the RTS pins indicates the transmitter has data ready and waiting to send. Writing a 1 in the modem control register (MCR[1]) sets these pins to low, indicating data is available. After a reset, these pins are set to 1. These pins only affect the transmit and receive operation when auto-RTS function is enabled through the enhanced feature register (EFR[6]), for hardware flow control operation.		
RXA, RXB,	5, 4	4, 3	I	Receive data input. These inputs are associated with individual serial channel data to the TL16C752C. During the local loopback mode, these RX input pins are disabled and TX data is internally connected to the UART RX input internally. During normal mode, RXn should be held high when no data is being received. These outputs also can be used in IrDA mode. See the IrDA mode section for more information.		
RXRDYA, RXRDYB	31, 18		0	Receive ready (active low). RXRDYA and RXRDYB go low when the trigger level has been reached or a timeout interrupt occurs. They go high when the RX FIFO is empty or there is an error in RX FIFO.		
TXA, TXB,	7, 8	5, 6	0	Transmit data. These outputs are associated with individual serial transmit channel data from the TL16C752C. During the local loopback mode, the TX input pin is disabled and TX data is internally connected to the UART RX input.		
TXRDYA, TXRDYB	43, 6		0	Transmit ready (active low). TXRDYA and TXRDYB go low when there are a trigger level number of spares available. They go high when the TX buffer is full.		
V _{CC}	42	26	Pwr	Power supply inputs		
XTAL1	13	9	I	Crystal or external clock input. XTAL1 functions as a crystal input or as an external clock input. A crystal can be connected between XTAL1 and XTAL2 to form an internal oscillator circuit (see Figure 10). Alternatively, an external clock can be connected to XTAL1 to provide custom data rates.		
XTAL2	14	10	0	Output of the crystal oscillator or buffered clock. See also XTAL1. XTAL2 is used as a crystal oscillator output or buffered clock output.		



FUNCTIONAL BLOCK DIAGRAM







A. The vote logic determines whether the RX data is a logic 1 or 0. It takes three samples of the RX line and uses a majority vote to determine the logic level received. The Vote logic operates on all bits received.

FUNCTIONAL DESCRIPTION

The TL16C752C UART is pin compatible with the TL16C2550 UART in the PFB package. It provides more enhanced features. All additional features are provided through a special enhanced feature register.

The UART performs serial-to-parallel conversion on data characters received from peripheral devices or modems and parallel-to-parallel conversion on data characters transmitted by the processor. The complete status of each channel of the TL16C752C UART can be read at any time during functional operation by the processor.

The TL16C752C UART can be placed in an alternate mode (FIFO mode) relieving the processor of excessive software overhead by buffering received/transmitted characters. Both the receiver and transmitter FIFOs can store up to 64 bytes (including three additional bits of error status per byte for the receiver FIFO) and have selectable or programmable trigger levels. Primary outputs RXRDY and TXRDY allow Signaling of DMA transfers.

The TL16C752C UART has selectable hardware flow control and software flow control. Both schemes significantly reduce software overhead and increase system efficiency by automatically controlling serial data flow. Hardware flow control uses the RTS output and CTS input signals. Software flow control uses programmable Xon/Xoff characters.

The TL16C752C includes a programmable baud rate generator that can divide the timing reference clock by a divisor between 1 and 65535. A bit (MCR7) can be used to invoke a pre-scaler (divide by 4) off the reference clock, prior to the baud rate generator input. The divide by 4 pre-scaler is selected when MCR7 is set to 1.

Trigger Levels

The TL16C752C UART provides independent selectable and programmable trigger levels for both receiver and transmitter DMA and interrupt generation. After reset, both transmitter and receiver FIFOs are disabled and so, in effect, the trigger level is the default value of one byte. The selectable trigger levels are available via the FCR. The programmable trigger levels are available via the TLR.

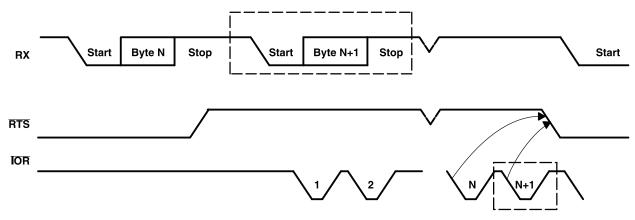
Hardware Flow Control

Hardware flow control is composed of auto- $\overline{\text{CTS}}$ and auto- $\overline{\text{RTS}}$. Auto- $\overline{\text{CTS}}$ and auto- $\overline{\text{RTS}}$ can be enabled/disabled independently by programming EFR[7:6].

With auto- $\overline{\text{CTS}}$, $\overline{\text{CTS}}$ must be active before the UART can transmit data. Auto- $\overline{\text{RTS}}$ only activates the $\overline{\text{RTS}}$ output when there is enough room in the FIFO to receive data and deactivates the $\overline{\text{RTS}}$ output when the RX FIFO is sufficiently full. The HALT and RESTORE trigger levels in the TCR determine the levels at which $\overline{\text{RTS}}$ is activated/deactivated. If both auto- $\overline{\text{CTS}}$ and auto- $\overline{\text{RTS}}$ are enabled, when $\overline{\text{RTS}}$ is connected to $\overline{\text{CTS}}$, data transmission does not occur unless the receiver FIFO has empty space. Thus, overrun errors are eliminated during hardware flow control. If not enabled, overrun errors occur if the transmit data rate exceeds the receive FIFO servicing latency.

Auto-RTS

Auto-RTS data flow control originates in the receiver block (see functional block diagram). Figure 1 shows RTS functional timing. The receiver FIFO trigger levels used in Auto-RTS are stored in the TCR. RTS is active if the RX FIFO level is below the HALT trigger level in TCR[3:0]. When the receiver FIFO HALT trigger level is reached, RTS is deasserted. The sending device (e.g., another UART) may send an additional byte after the trigger level is reached (assuming the sending UART has another byte to send) because it may not recognize the deassertion of RTS until it has begun sending the additional byte. RTS is automatically reasserted once the receiver FIFO reaches the RESUME trigger level programmed via TCR[7:4]. This reassertion allows the sending device to resume transmission.



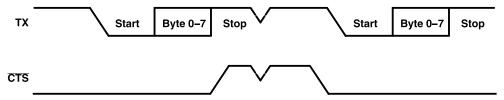
- A. N = receiver FIFO trigger level B.
- B. The two blocks in dashed lines cover the case where an additional byte is sent as described in Auto-RTS.

Figure 1. RTS Functional Timing

Auto-CTS

The transmitter circuitry checks $\overline{\text{CTS}}$ before sending the next data byte. When $\overline{\text{CTS}}$ is active, the transmitter sends the next byte. To stop the transmitter from sending the following byte, $\overline{\text{CTS}}$ must be deasserted before the middle of the last stop bit that is currently being sent. The auto- $\overline{\text{CTS}}$ function reduces interrupts to the host system. When flow control is enabled, the $\overline{\text{CTS}}$ state changes and need not trigger host interrupts because the device automatically controls its own transmitter. Without auto- $\overline{\text{CTS}}$, the transmitter sends any data present in the transmit FIFO and a receiver overrun error can result. Figure 2 shows $\overline{\text{CTS}}$ functional timing, and Figure 3 shows an example of autoflow control.





- A. When $\overline{\text{CTS}}$ is low, the transmitter keeps sending serial data out.
- B. When CTS goes high before the middle of the last stop bit of the current byte, the transmitter finishes sending the current byte, but it does not send the next byte.
- C. When CTS goes from high to low, the transmitter begins sending data again.

Figure 2. CTS Functional Timing

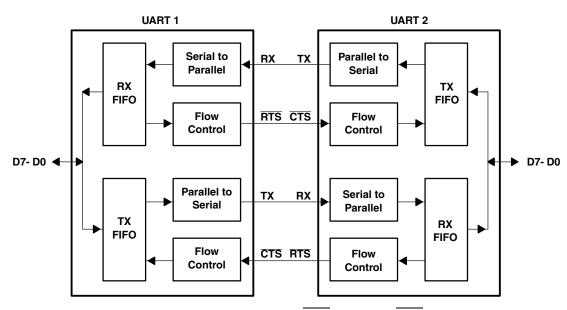


Figure 3. Autoflow Control (Auto-RTS and Auto-CTS) Example

Software Flow Control

Software flow control is enabled through the enhanced feature register and the modem control register. Different combinations of software flow control can be enabled by setting different combinations of EFR[3-0]. Table 1 shows software flow control options.

Two other enhanced features relate to S/W flow control:

 Xon Any Function [MCR(5): Operation resumes after receiving any character after recognizing the Xoff character.

NOTE

It is possible that an Xon1 character is recognized as an Xon Any character, which could cause an Xon2 character to be written to the RX FIFO.

• **Special Character [EFR(5)]**: Incoming data is compared to Xoff2. Detection of the special character sets the Xoff interrupt {IIR(4)] but does not halt transmission. The Xoff interrupt is cleared by a read of the IIR. The special character is transferred to the RX FIFO.



Table 1. Software Flow Control Options EFR[3:0]

BIT 3	BIT 2	BIT 1	BIT 0	Tx, Rx SOFTWARE FLOW CONTROLS
0	0	Х	Х	No transmit flow control
1	0	Х	Х	Transmit Xon1, Xoff1
0	1	Х	Х	Transmit Xon2, Xoff2
1	1	Х	Х	Transmit Xon1, Xon2: Xoff1, Xoff2
Х	Х	0	0	No receive flow control
Х	Х	1	0	Receiver compares Xon1, Xoff1 X X 0 1
Х	Х	0	1	Receiver compares Xon2, Xoff2
1	0	1	1	Transmit Xon1, Xoff1 Receiver compares Xon1 or Xon2, Xoff1 or Xoff2
0	1	1	1	Transmit Xon2, Xoff2 Receiver compares Xon1 or Xon2, Xoff1 or Xoff2
1	1	1	1	Transmit Xon1, Xon2: Xoff1, Xoff2 Receiver compares Xon1 and Xon2: Xoff1 and Xoff2
0	0	1	1	No transmit flow control Receiver compares Xon1 and Xon2: Xoff1 and Xoff2

When software flow control operation is enabled, the TL16C752C compares incoming data with Xoff1/2 programmed characters (in certain cases Xoff1 and Xoff2 must be received sequentially⁽¹⁾). When an Xoff character is received, transmission is halted after completing transmission of the current character. Xoff character detection also sets IIR[4] and causes INT to go high (if enabled via IER[5]).

To resume transmission an Xon1/2 character must be received (in certain cases Xon1 and Xon2 must be received sequentially). When the correct Xon characters are received IIR[4] is cleared and the Xoff interrupt disappears.

NOTE

If a parity, framing or break error occurs while receiving a software flow control character, this character is treated as normal data and is written to the RCV FIFO.

Xoff1/2 characters are transmitted when the RX FIFO has passed the programmed trigger level TCR[3:0].

Xon1/2 characters are transmitted when the RX FIFO reaches the trigger level programmed via TCR[7:4].

NOTE

If, after an Xoff character has been sent, software flow control is disabled, the UART transmits Xon characters automatically to enable normal transmission to proceed. A feature of the TL16C752C UART design is that if the software flow combination (EFR[3:0]) changes after an Xoff has been sent, the originally programmed Xon is automatically sent. If the RX FIFO is still above the trigger level the newly programmed Xoff1/2 is transmitted.

The transmission of Xoff/Xon(s) follows the exact same protocol as transmission of an ordinary byte from the FIFO. This means that even if the word length is set to be 5, 6, or 7 characters, then the 5, 6, or 7 least significant bits of Xoff1,2/Xon1,2 are transmitted. The transmission of 5, 6, or 7 bits of a character is seldom done, but this functionality is included to maintain compatibility with earlier designs.

It is assumed that software flow control and hardware flow control are never enabled simultaneously. Figure 4 shows a software flow control example.

⁽¹⁾ When pairs of Xon/Xoff characters are programmed to occur sequentially, received Xon1/Xoff1 characters will be written to the Rx FIFO if the subsequent character is not Xon2/Xoff2.



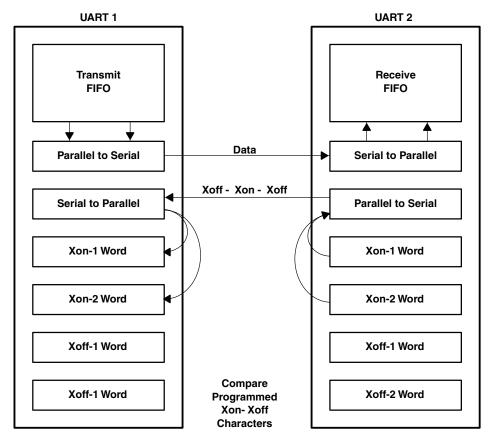


Figure 4. Software Flow Control Example

Software Flow Control Example

Assumptions: UART1 is transmitting a large text file to UART2. Both UARTs are using software flow control with single character Xoff (0F) and Xon (0D) tokens. Both have Xoff threshold (TCR [3:0]=F) set to 60 and Xon threshold (TCR[7:4]=B) set to 32. Both have the interrupt receive threshold (TLR[7:4]=D) set to 52.

UART1 begins transmission and sends 52 characters, at which point UART2 generates an interrupt to its processor to service the RCV FIFO, but assumes the interrupt latency is fairly long. UART1 continues sending characters until a total of 60 characters have been sent. At this time UART2 transmits a 0F to UART1, informing UART1 to halt transmission. UART1 will likely send the 61st character while UART2 is sending the Xoff character. Now UART2 is serviced and the processor reads enough data out of the RCV FIFO that the level drops to 32. UART2 now sends a 0D to UART1, informing UART1 to resume transmission.

Reset

Table 2 summarizes the state of outputs after reset.



Table 2. Register Reset Functions⁽¹⁾

REGISTER	RESET CONTROL	RESET STATE
Interrupt enable register	RESET	All bits cleared
Interrupt identification register	RESET	Bit 0 is set. All other bits cleared.
FIFO control register	RESET	All bits cleared
Line control register	RESET	Reset to 00011101 (1D hex).
Modem control register	RESET	All bits cleared
Line status register	RESET	Bits 5 and 6 set. All other bits cleared.
Modem status register	RESET	Bits 0–3 cleared. Bits 4–7 input signals.
Enhanced feature register	RESET	All bits cleared
Receiver holding register	RESET	Pointer logic cleared
Transmitter holding register	RESET	Pointer logic cleared
Transmission control register	RESET	All bits cleared
Trigger level register	RESET	All bits cleared
Alternate function register	RESET	All bits (except AFR4) cleared; AFR4 set

⁽¹⁾ Registers DLL, DLH, SPR, Xon1, Xon2, Xoff1, Xoff2 are not reset by the top-level reset signal RESET, i.e., they hold their initialization values during reset.

Table 3 summarizes the state of outputs after reset.

Table 3. Signal Reset Functions

SIGNAL	RESET CONTROL	RESET STATE
TX	RESET	High
RTS	RESET	High
DTR	RESET	High
RXRDYA-B	RESET	High
TXRDYA-B	RESET	Low

Interrupts

The TL16C752C UART has interrupt generation and prioritization (six prioritized levels of interrupts) capability. The interrupt enable register (IER) enables each of the six types of interrupts and the INT signal in response to an interrupt generation. The IER also can disable the interrupt system by clearing bits 0–3, 5–7. When an interrupt is generated, the interrupt identification register(IIR) indicates that an interrupt is pending and provides the type of interrupt through IIR[5–0]. Table 4 summarizes the interrupt control functions.

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Table 4. Interrupt Control Functions

IIR[5-0]	PRIORITY LEVEL	INTERRUPT TYPE	INTERRUPT SOURCE	INTERRUPT RESET METHOD
000001	None	None	None	None
000110	1	Receiver line status	OE, FE, PE, or BI errors occur in characters in the RX FIFO	FE < PE < BI: All erroneous characters are read from the RX FIFO. OE: Read LSR
001100	2	RX timeout	Stale data in RX FIFO	Read RHR
000100	2	RHR interrupt	DRDY (data ready) (FIFO disable) RX FIFO above trigger level (FIFO enable)	Read RHR
000010	3	THR interrupt	TFE (THR empty) (FIFO disable) TX FIFO passes above trigger level (FIFO enable)	Read IIR or a write to the THR
000000	4	Modem status	MSR[3:0]= 0	Read MSR
010000	5	Xoff interrupt	Receive Xoff character(s)/special character	Receive Xon character(s)/Read of IIR
100000	6	CTS, RTS	RTS pin or CTS pin change state from active (low) to inactive (high)	Read IIR

It is important to note that for the framing error, parity error, and break conditions, LSR[7] generates the interrupt. LSR[7] is set when there is an error anywhere in the RX FIFO and is cleared only when there are no more errors remaining in the FIFO. LSR[4–2] always represent the error status for the received character at the top of the Rx FIFO. Reading the Rx FIFO updates LSR[4–2] to the appropriate status for the new character at the top of the FIFO. If the Rx FIFO is empty, then LSR[4–2] is all zeros.

For the Xoff interrupt, if an Xoff flow character detection caused the interrupt, the interrupt is cleared by an Xon flow character detection. If a special character detection caused the interrupt, the interrupt is cleared by a read of the ISR.

Interrupt Mode Operation

In interrupt mode (if any bit of IER[3:0] is1), the processor is informed of the status of the receiver and transmitter by an interrupt signal, INT. Therefore, it is not necessary to continuously poll the line status register (LSR) to see if any interrupt needs to be serviced. Figure 5 shows interrupt mode operation.

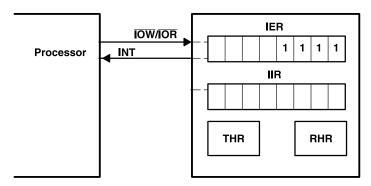


Figure 5. Interrupt Mode Operation

Polled Mode Operation

In polled mode (IER[3:0] = 0000), the status of the receiver and transmitter can then be checked by polling the line status register (LSR). This mode is an alternative to the interrupt mode of operation where the status of the receiver and transmitter is automatically known by means of interrupts sent to the CPU. Figure 6 shows polled mode operation.



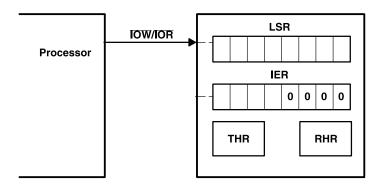


Figure 6. FIFO Polled Mode Operation

DMA Signaling

There are two modes of DMA operation, DMA mode 0 or 1, selected by FCR[3].

In DMA mode 0 or FIFO disable (FCR[0]=0) DMA occurs in single character transfers. In DMA mode 1 multicharacter (or block) DMA transfers are managed to relieve the processor for longer periods of time.

Single DMA Transfers (DMA Mode0/FIFO Disable)

Transmitter: When empty, the \overline{TXRDY} signal becomes active. \overline{TXRDY} goes inactive after one character has been loaded into it.

Receiver: RXRDY is active when there is at least one character in the FIFO. It becomes inactive when the receiver is empty.

Figure 7 shows TXRDY and RXRDY in DMA mode 0/FIFO disable.

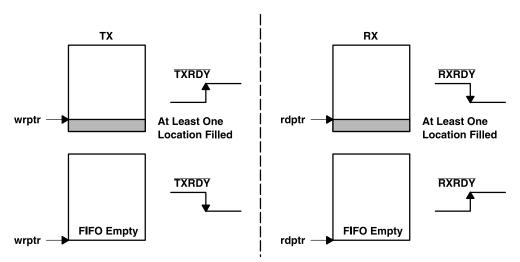


Figure 7. TXRDY and RXRDY in DMA Mode 0/FIFO Disable

Block DMA Transfers (DMA Mode1)

Transmitter: TXRDY is active when a trigger level number of spaces are available. It becomes inactive when the FIFO is full.

Receiver: RXRDY becomes active when the trigger level has been reached or when a timeout interrupt occurs. It goes inactive when the FIFO is empty or an error in the RX FIFO is flagged by LSR(7).

Figure 8 shows TXRDY and RXRDY in DMA mode 1.



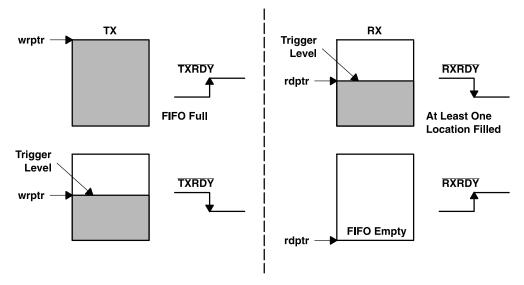


Figure 8. TXRDY and RXRDY in DMA Mode 1

Sleep Mode

Sleep mode is an enhanced feature of the TL16C752C UART. It is enabled when EFR[4], the enhanced functions bit, is set and when IER[4] is set. Sleep mode is entered when:

- The serial data input line, RX, is idle (see break and time-out conditions).
- The TX FIFO and TX shift register are empty.
- There are no interrupts pending except THR and timeout interrupts.

Sleep mode is not entered if there is data in the RX FIFO.

In sleep mode the UART clock and baud rate clock are stopped. Because most registers are clocked using these clocks the power consumption is greatly reduced. The UART wakes up when any change is detected on the RX line, when there is any change in the state of the modem input pins or if data is written to the TX FIFO.

NOTE

Writing to the divisor latches, DLL and DLH, to set the baud clock, must not be done during sleep mode. Therefore it is advisable to disable sleep mode using IER[4] before writing to DLL or DLH.

Break and Timeout Conditions

An RX timeout condition is detected when the receiver line, RX, has been high for a time equivalent to $(4 \times programmed word length) + 12$ bits and there is at least one byte stored in the Rx FIFO.

When a break condition occurs, the TX line is pulled low. A break condition is activated by setting LCR[6].

Programmable Baud Rate Generator

The TL16C752C UART contains a programmable baud generator that divides reference clock by a divisor in the range between 1 and (216–1). The output frequency of the baud rate generator is 16x the baud rate. An additional divide-by-4 prescaler is also available and can be selected by MCR[7], as shown in the following. The formula for the divisor is:

Divisor = (XTAL crystal input frequency / prescaler) / (desired baud rate x 16)

Where

```
prescaler = 

1 when CLKSEL = high during reset, or MCR[7] is set to 0 after reset
4 when CLKSEL = high during reset, or MCR[7] is set to 1 after reset
```

Figure 9 shows the internal prescaler and baud rate generator circuitry.

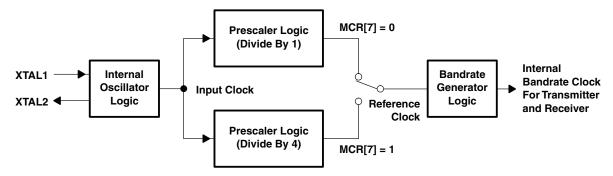


Figure 9. Prescaler and Baud Rate Generator Block Diagram

DLL and DLH must be written to in order to program the baud rate. DLL and DLH are the least significant and most significant byte of the baud rate divisor. If DLL and DLH are both zero, the UART is effectively disabled, as no baud clock is generated. The programmable baud rate generator is provided to select both the transmit and receive clock rates. Table 5 and Table 6 show the baud rate and divisor correlation for the crystal with frequency 1.8432 MHz and 3.072 MHz, respectively.

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Table 5. Baud Rates Using a 1.8432-MHz Crystal

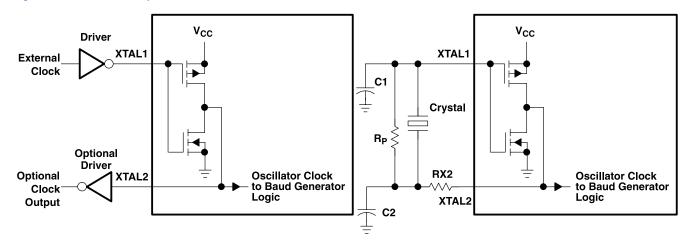
DESIRED BAUD RATE	DIVISOR USED TO GENERATE 16× CLOCK	PERCENT ERROR DIFFERENCE BETWEEN DESIRED AND ACTUAL
50	2304	
75	1536	
110	1047	0.026
134.5	857	0.058
150	768	
300	384	
600	192	
1200	96	
1800	64	
2000	58	0.69
2400	48	
3600	32	
4800	24	
7200	16	
9600	12	
19200	6	
38400	3	
56000	2	2.86

Table 6. Baud Rates Using a 3.072-MHz Crystal

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DESIRED BAUD RATE	DIVISOR USED TO GENERATE 16× CLOCK	PERCENT ERROR DIFFERENCE BETWEEN DESIRED AND ACTUAL				
50	3840					
75	2560					
110	1745	0.026				
134.5	1428	0.034				
150	1280					
300	640					
600	320					
1200	160					
1800	107	0.312				
2000	96					
2400	80					
3600	53	0.628				
4800	40					
7200	27	1.23				
9600	20					
19200	10					
38400	5					



Figure 10 shows the crystal clock circuit reference.



TYPICAL CRYSTAL OSCILLATOR NETWORK

CRYSTAL	R _P	RX2	C1	C2
16 MHz	1 ΜΩ	0	33 pF	33 pF
3.072 MHz	1 ΜΩ	1.5 kΩ	10-30 pF	40-60 pF
1.8432 MHz	1 ΜΩ	1.5 kΩ	10-30 pF	40-60 pF

- A. For crystal with fundamental frequency from 1 MHz to 24 MHz
- B. For input clock frequency higher then 24 MHz, the crystal is not allowed and the oscillator must be used, because the TL16C752C internal oscillator cell can only support the crystal frequency up to 24 MHz.

Figure 10. Typical Crystal Clock Circuits

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ABSOLUTE MAXIMUM RATINGS(1)

over operating free-air temperature range (unless otherwise noted)

	PARAME	TER	MIN	MAX	UNIT
V _{CC}	Supply voltage range	Supply voltage range			V
VI	Input voltage range			$V_{CC} + 0.5$	V
Vo	Output voltage range		-0.5	V _{CC} + 0.5	V
_	O	TL16C752C	0	70	°C
T _A	Operating free-air temperature range	ng free-air temperature range TL16C752CI	-40	85	٠.
T _{stg}	Storage temperature range		-65	150	°C

⁽¹⁾ Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

Table 7. TYPICAL PACKAGE THERMAL RESISTANCE DATA

PACKAGE		
32-pin QFN RHB	$\theta_{JA} = xx^{\circ}C/W$	$\theta_{JC} = xx^{\circ}C/W$
48-pin TQFP PFB	$\theta_{JA} = xx^{\circ}C/W$	$\theta_{JC} = xx^{\circ}C/W$

Table 8. TYPICAL PACKAGE WEIGHT

PACKAGE	WEIGHT IN GRAMS
32-pin QFN RHB	0.25
48-pin TQFP PFB	0.30

RECOMMENDED OPERATING CONDITIONS, $V_{CC} = 1.8 \text{ V} \pm 10\%$

over operating free-air temperature range (unless otherwise noted)

	PARAM	ETER	MIN	NOM	MAX	UNIT
V_{CC}	Supply voltage		1.62	1.8	1.98	V
VI	Input voltage	0		V _{CC}	V	
V_{IH}	High-level input voltage	1.4		1.98	V	
V _{IL}	Low-level input voltage	-0.3		0.4	V	
Vo	Output voltage		0		V _{CC}	V
I _{OH}	High-level output current	All outputs			0.5	mA
I_{OL}	Low-level output current	All outputs			1	mA
	Oscillator/clock speed			16	MHz	

RECOMMENDED OPERATING CONDITIONS, $V_{CC} = 2.5 \text{ V} \pm 10\%$

over operating free-air temperature range (unless otherwise noted)

	PARAM	MIN	NOM	MAX	UNIT	
V_{CC}	Supply voltage		2.25	2.5	2.75	V
VI	Input voltage	0		V _{CC}	V	
V _{IH}	High-level input voltage	1.8		2.75	V	
V _{IL}	Low-level input voltage	-0.3		0.6	V	
Vo	Output voltage		0		V_{CC}	V
I _{OH}	High-level output current	All outputs			1	mA
I _{OL}	Low-level output current	All outputs			2	mA
	Oscillator/clock speed			24	MHz	

RECOMMENDED OPERATING CONDITIONS, $V_{CC} = 3.3 \text{ V} \pm 10\%$

over operating free-air temperature range (unless otherwise noted)

	PARAM	ETER	MIN	NOM	MAX	UNIT
V _{CC}	Supply voltage		3	3.3	3.6	V
VI	Input voltage	Input voltage				V
V _{IH}	High-level input voltage	0.7 × V _{CC}			V	
V _{IL}	Low-level input voltage			0.3 × V _{CC}	V	
Vo	Output voltage		0		V _{CC}	V
I _{OH}	High-level output current	All outputs			1.8	mA
I _{OL}	Low-level output current	All outputs			3.2	mA
	Oscillator/clock speed			32	MHz	

RECOMMENDED OPERATING CONDITIONS, V_{CC} = 5 V ±10%

over operating free-air temperature range (unless otherwise noted)

	PARAM	IETER	MIN N	OM MAX	UNIT
V_{CC}	Supply voltage			5.5	V
VI	Input voltage		V _{CC}	V	
V	High level input value of	Except XIN	0		V
V _{IH}	High-level input voltage	XIN	0.7 × V _{CC}		
		Except XIN		0.8	
V_{IL}	Low-level input voltage	XIN		0.3 × V _{CC}	V
Vo	Output voltage		0	V _{CC}	V
I _{OH}	High-level output current	All outputs		4	mA
I _{OL}	Low-level output current	All outputs		4	mA
	Oscillator/clock speed			48	MHz

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ELECTRICAL CHARACTERISTICS, V_{CC} = 1.8 V

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CO	NDITIONS	MIN	TYP	MAX	UNIT
V _{OH}	High-level output voltage	$I_{OH} = -0.5 \text{ mA}$		1.3			V
V _{OL}	Low-level output voltage	I _{OL} = 1 mA				0.5	V
I _I	Input current	V _{CC} = 1.98 V, V _I = 0 to 1.98 V,	V _{SS} = 0, All other terminals floating			10	μΑ
High-impedance state output current		$V_{CC} = 1.98 \text{ V},$ $V_{O} = 0 \text{ to } 1.98 \text{ V},$, ,				μA
	output ourrent	Chip selected in write mode or ch					
	Supply current	$V_{CC} = 1.98 V,$	$T_A = 0$ °C,				
I _{CC}		SIN, $\overline{\text{DSR}}$, $\overline{\text{DCD}}$, $\overline{\text{CTS}}$, and $\overline{\text{RI}}$ at			4.5	mA	
		All other inputs at 0.4 V, No load on outputs,	XTAL1 at 16 MHz, Baud rate = 1 Mbit/s	1		1.0	IIIA
C _{I(CLK)}	Clock input capacitance				5	7	рF
C _{O(CLK)}	Clock output capacitance	$V_{CC} = 0$	$V_{SS} = 0$,		5	7	рF
C _I	Input capacitance	f = 1 MHz, All other terminals grounded	$T_A = 25^{\circ}C$,		6	10	рF
Co	Output capacitance				10	15	pF

ELECTRICAL CHARACTERISTICS, V_{CC} = 2.5 V

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CO	NDITIONS	MIN	TYP	MAX	UNIT
V _{OH}	High-level output voltage	I _{OH} = -1 mA		1.8			V
V _{OL}	Low-level output voltage	I _{OL} = 2 mA				0.5	V
II	Input current	$V_{CC} = 2.75 \text{ V},$ $V_{I} = 0 \text{ to } 2.75 \text{ V},$	V _{SS} = 0, All other terminals floating			10	μΑ
l _{OZ}	High-impedance state output current	$V_{CC} = 2.75 \text{ V},$ $V_{O} = 0 \text{ to } 2.75 \text{ V},$ Chip selected in write mode or compared to the selected of the selecte	$V_{SS} = 0$,			±20	μΑ
	Supply ourrent	$V_{CC} = 2.75 \text{ V},$ SIN, \overline{DSR} , \overline{DCD} , \overline{CTS} , and \overline{RI} at			90	mA	
I _{CC}	Supply current	All other inputs at 0.6 V, No load on outputs,	XTAL1 at 24 MHz, Baud rate = 1.5 Mbit/s			90	ША
C _{I(CLK)}	Clock input capacitance				5	7	pF
C _{O(CLK)}	Clock output capacitance	V _{CC} = 0,	$V_{SS} = 0$,		5	7	pF
C _I	Input capacitance	f = 1 MHz, All other terminals grounded	$V_{SS} = 0,$ $T_A = 25$ °C,		6	10	pF
Co	Output capacitance				10	15	pF

ELECTRICAL CHARACTERISTICS, V_{CC} = 3.3 V

over operating free-air temperature range (unless otherwise noted)

	PARAMETER	TEST CO	ONDITIONS	MIN	TYP	MAX	UNIT
V _{OH}	High-level output voltage	I _{OH} = −1.8 mA		2.4			V
V _{OL}	Low-level output voltage	I _{OL} = 3.2 mA				0.5	V
I _I	Input current	$V_{CC} = 3.6 \text{ V},$ $V_{I} = 0 \text{ to } 3.6 \text{ V},$	V _{SS} = 0, All other terminals floating			10	μΑ
High-impedance state output current		V _{CC} = 3.6 V, V _O = 0 to 3.6 V,					μA
	output current	Chip selected in write mode or					
		$V_{CC} = 3.6 \text{ V},$	$T_A = 0$ °C,				
I _{CC}	Supply current	SIN, $\overline{\text{DSR}}$, $\overline{\text{DCD}}$, $\overline{\text{CTS}}$, and $\overline{\text{RI}}$ a			16	mA	
100	Зирріу сипені	All other inputs at 0.8 V, No load on outputs,	XTAL1 at 32 MHz, Baud rate = 2 Mbit/s			10	1117
C _{I(CLK)}	Clock input capacitance				5	7	pF
C _{O(CLK)}	Clock output capacitance	$V_{CC} = 0$,	$V_{SS} = 0$,		5	7	pF
Cı	Input capacitance	f = 1 MHz, All other terminals grounded	$T_A = 25^{\circ}C$,		6	10	pF
Co	Output capacitance				10	15	pF

ELECTRICAL CHARACTERISTICS, V_{CC} = 5 V

over operating free-air temperature range (unless otherwise noted)

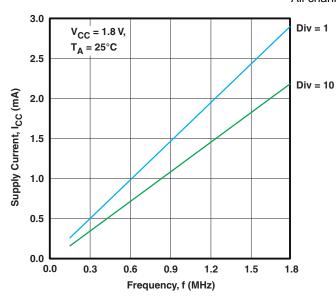
	PARAMETER	TEST CO	MIN	TYP	MAX	UNIT	
V _{OH}	High-level output voltage	$I_{OH} = -4 \text{ mA}$		4			V
V _{OL}	Low-level output voltage	I _{OL} = 4 mA				0.4	V
I _I	Input current	V _{CC} = 5.5 V, V _I = 0 to 5.5 V,	V _{SS} = 0, All other terminals floating			10	μΑ
I _{OZ}	High-impedance state output current	$V_{CC} = 5.5 \text{ V},$ $V_{O} = 0 \text{ to } 5.5 \text{ V},$	00 /			±20	μΑ
		Chip selected in write mode or					
		$V_{CC} = 5.5 \text{ V}, T_{A} = 0^{\circ}\text{C},$					
I _{CC}	Supply current	SIN, $\overline{\text{DSR}}$, $\overline{\text{DCD}}$, $\overline{\text{CTS}}$, and $\overline{\text{RI}}$ a			40	mA	
icc	Сарріу балолі	All other inputs at 0.8 V, No load on outputs,	XTAL1 at 48 MHz, Baud rate = 3 Mbit/s			40	1117
C _{I(CLK)}	Clock input capacitance				5	7	pF
C _{O(CLK)}	Clock output capacitance	$V_{CC} = 0$	$V_{SS} = 0$,		5	7	pF
Cı	Input capacitance	f = 1 MHz, All other terminals grounded	$T_A = 25^{\circ}C$,		6	10	pF
Co	Output capacitance				10	15	pF

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TYPICAL CHARACTERISTICS

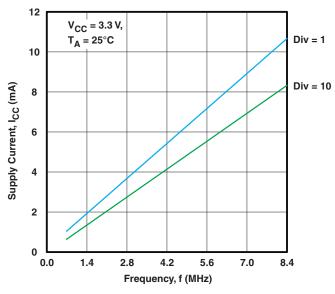
All channels active



Div = 1 $V_{CC} = 2.5 V$, T_A = 25°C 5 Div = 10 Supply Current, I_{CC} (mA) 4 3 2 1 0.7 2.1 2.8 3.5 0.0 4.2 Frequency, f (MHz)

Figure 11. Supply Current vs Frequency ($V_{CC} = 1.8 \text{ V}$)

Figure 12. Supply Current vs Frequency (V_{CC} = 2.5 V)



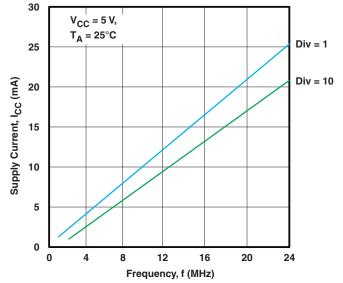


Figure 13. Supply Current vs Frequency (V_{CC} = 3.3 V)

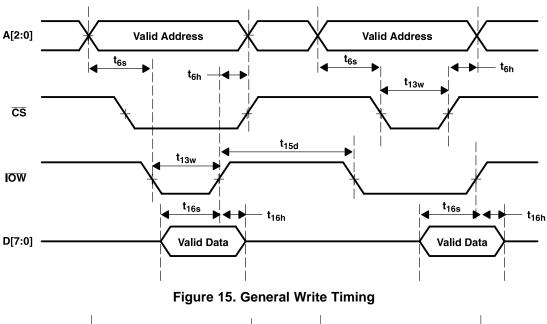
Figure 14. Supply Current vs Frequency (V_{CC} = 5 V)

TIMING REQUIREMENTS

 $T_A = 0$ °C to 70°C, $V_{CC} = 1.8$ V to 5 V ±10% (unless otherwise noted)

						LIM	ITS				
	PARAMETER	TEST CONDITIONS	1.8	V	2.5	5 V	3.3	3 V	5	٧	UNIT
		CONDITIONS	MIN	MAX	MIN	MAX	MIN	MAX	MIN	MAX	•
t _{RES} ET	Reset pulse width		200		200		200		200		ns
C_{P}	CP Clock period		63		42		32		20		ns
t _{3w}	Oscillator/Clock speed			16		24		32		48	MHz
t _{6s}	Address setup time		20		15		10		5		ns
t _{6h}	Address hold time	See Figure 15 and Figure 16	15		10		7		5		ns
t _{7w}	IOR strobe width	See Figure 15 and Figure 16	85		70		50		40		ns
t _{9d}	Read cycle delay	See Figure 16	85		70		60		50		ns
t _{12d}	Delay from IOR to data	See Figure 16		65		50		35		25	ns
t _{12h}	Data disable time			35		25		20		15	ns
t _{13w}	IOW strobe width	See Figure 15	85		70		50		40		ns
t _{15d}	Write cycle delay	See Figure 15	85		70		60		50		ns
t _{16s}	Data setup time	See Figure 15	40		30		20		15		ns
t _{16h}	Data hold time	See Figure 15	35		25		15		10		ns
t _{17d}	Delay from IOW to output	50 pF load, See Figure 17		60		40		30		20	ns
t _{18d}	Delay to set interrupt from MODEM input	50 pF load, See Figure 17		70		55		45		35	ns
t _{19d}	Delay to reset interrupt from IOR	50 pF load		80		55		40		30	ns
t _{20d}	Delay from stop to set interrupt	See Figure 18		1		1		1		1	Baudrate
t _{21d}	Delay from IOR to reset interrupt	50 pF load, See Figure 18		55		45		35		25	ns
t _{22d}	Delay from stop to interrupt	See Figure 21		1		1		1		1	Baudrate
t _{23d}	Delay from initial IOW reset to transmit star	See Figure 21	8	24	8	24	8	24	8	24	Baudrate
t _{24d}	Delay from IOW to reset interrupt	See Figure 21		75		45		35		25	ns
t _{25d}	Delay from stop to set RXRDY	See Figure 19 and Figure 20		1		1		1		1	Baudrate
t _{26d}	Delay from IOR to reset RXRDY	See Figure 19 and Figure 20		1		1		1		1	μs
t _{27d}	Delay from IOW to set TXRDY	See Figure 22 and Figure 23		70		60		50		40	ns
t _{28d}	Delay from start to reset TXRDY	See Figure 22 and Figure 23		16		16		16		16	Baudrate





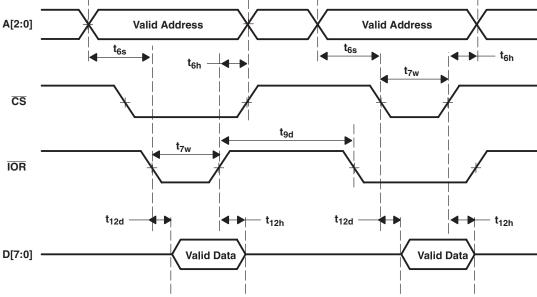


Figure 16. General Read Timing

www.ti.com SLLS646A - MARCH 2008 - REVISED AUGUST 2009 Active IOW t_{17d} RTS (A-B) Change of State DTR (A-B) **Change of State** CD (A-B) **Change of State** CTS (A-B) DSR (A-B) t_{18d} INT (A-B) Active Active Active **←** t_{19d} Active Active Active IOR t_{18d} **Change of State** RI (A-B) Figure 17. Modem/Output Timing Start Stop Bit Bit Data Bits (5-8) D2 D3 D4 D5 D6 D7 RX (A-B) **Parity** Next 5 Data Bits Bit Data 6 Data Bits Start 7 Data Bits Bit - t_{20d} INT (A-B) Active

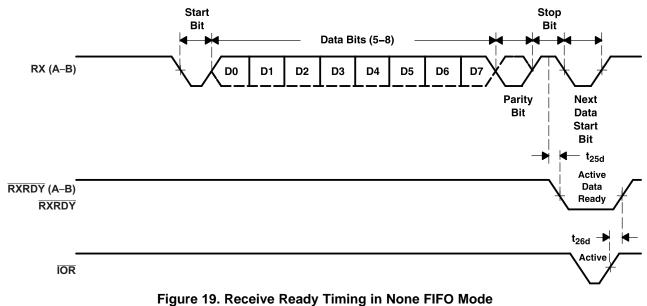
Figure 18. Receive Timing

16-Baud Rate Clock

IOR

t_{21d}





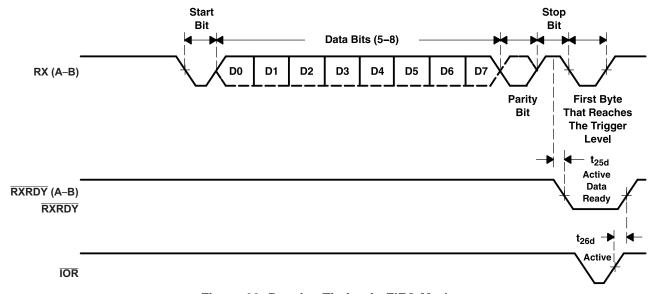


Figure 20. Receive Timing in FIFO Mode



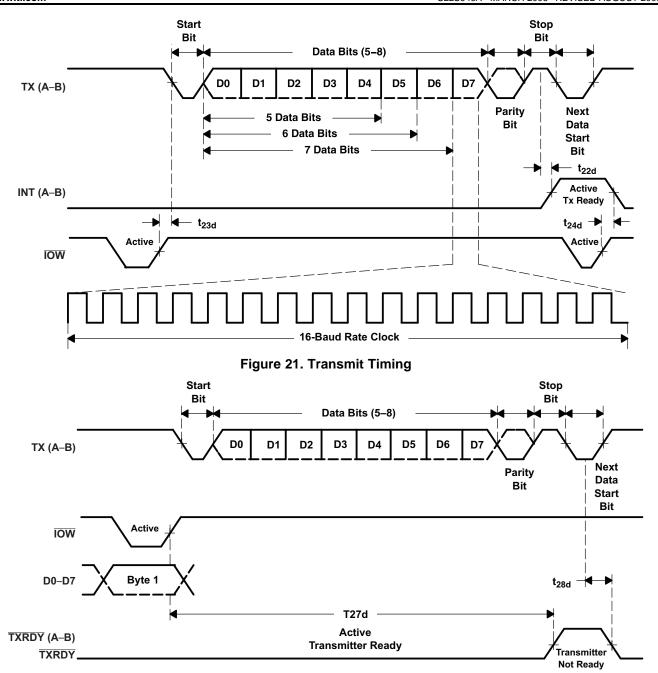


Figure 22. Transmit Ready Timing in None FIFO Mode



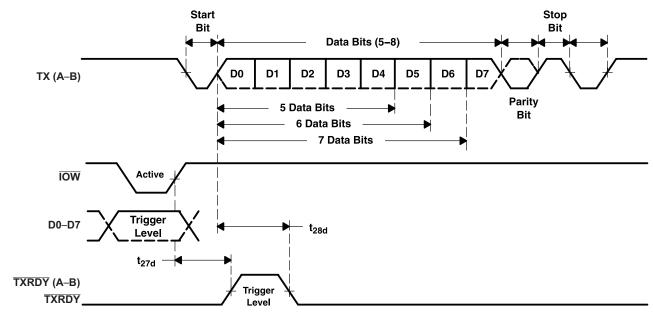


Figure 23. Transmit Timing in FIFO Mode

PRINCIPLES OF OPERATION

Register Map

Each register is selected using address lines A[0], A[1], A[2] and, in some cases, bits from other registers. The programming combinations for register selection are shown in Table 9.

Table 9. Register Map – Read/Write Properties (1)

A[2]	A[1]	A[0]	READ MODE	WRITE MODE
0	0	0	Receive holding register (RHR)	Transmit holding register (THR)
0	0	1	Interrupt enable register (IER)	Interrupt enable register
0	1	0	Interrupt identification register (IIR)	FIFO control register (FCR)
0	1	1	Line control register (LCR)	Line control register
1	0	0	Modem control register (MCR)	Modem control register
1	0	1	Line status register (LSR)	
1	1	0	Modem status register (MSR)	
1	1	1	Scratch register (SPR)	Scratch register (SPR)
0	0	0	Divisor latch LSB (DLL)	Divisor latch LSB (DLL)
0	0	1	Divisor latch MSB (DLH)	Divisor latch MSB (DLH)
0	1	0	Alternate function register (AFR)	Alternate function register (AFR)
0	1	0	Enhanced feature register (EFR)	Enhanced feature register
1	0	0	Xon-1 word	Xon-1 word
1	0	1	Xon-2 word	Xon-2 word
1	1	0	Xoff-1 word	Xoff-1 word
1	1	1	Xoff-2 word	Xoff-2 word
1	1	0	Transmission control register (TCR)	Transmission control register
1	1	1	Trigger level register (TLR)	Trigger level register
1	1	1	FIFO ready register	

⁽¹⁾ DLL and DLH are accessible only when LCR bit 7 is 1, and AFR is only accessible when LCR[7:5] = 100. Enhanced feature register, Xon1, 2 and Xoff1, 2 are accessible only when LCR is set to 10111111 (8hBF). Transmission control register and trigger level register are accessible only when EFR[4] = 1 and MCR[6] = 1, i.e. EFR[4] and MCR[6] are read/write enables. FCR FIFORdy register is accessible when any CS A-D = 0, MCR[2] = 1 and loopback MCR [4] = 0 is disabled.

MCR[7] can only be modified when EFR[4] is set.

Table 10 lists and describes the TL16C752C internal registers.



Table 10. TL16C752C Internal Registers (1) (2)

SPECIAL CONSIDERATIONS	ADDR	REGISTER	BIT 7	BIT 6	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	READ/ WRITE
	000	RHR	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit1	bit 0	Read
	000	THR	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit1	bit 0	Write
LCR[7] = 0	001	IER	CTS interrupt enable	RTS interrupt enable	Xoff interrupt enable	Sleep mode	Modem status interrupt	Rx line status interrupt	THR empty interrupt	Rx data available interrupt	Read/ write
	010	FCR	Rx trigger level	Rx trigger level	TX trigger level	TX trigger level	DMA mode select	Resets Tx FIFO	Resets Rx FIFO	Enables FIFOs	Write
	010	IIR	FCR(0)	FCR(0)	CTS, RTS	Xoff	Interrupt priority Bit 2	Interrupt priority Bit 1	Interrupt priority Bit 0	Interrupt status	Read
None	011	LCR	DLAB and EFR enable	Break control bit	Sets parity	Parity type select	Parity enable	No. of stop bits	Word length	Word length	Read/ write
	100	MCR	1x or 4x clock	TCR and TLR enable	Xon any	Enable loopback	IRQ enable	FIFORdy Enable	RTS	DTR	Read/ write
LCR[7:0] ≠ 1011 1111	101	LSR	Error in Rx FIFO	THR and TSR empty	THR empty	Break interrupt	Framing error	Parity error	Overrun error	Data in receiver	Read
	110	MSR	CD	RI	DSR	CTS	Δ CD	ΔRI	ΔDSR	ΔCTS	Read
	111	SPR	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit1	bit 0	Read/ write
L CD[7] 4	000	DLL	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit1	bit 0	Read/ write
LCR[7] = 1	001	DLH	bit 15	bit 14	bit 13	bit 12	bit 11	bit 10	bit 9	bit 8	Read/ write
LCR[7:5] = 100	010	AFR	DLY2	DLY1	DLY0	RCVEN	485LG	485RN	IREN	CONC	Read/ write
	010	EFR	Auto-CTS	Auto-RTS	Special character detect	Enable enhanced-functions	S/W flow control Bit 3	S/W flow control Bit 2	S/W flow control Bit 1	S/W flow control Bit 0	Read/ write
	100	Xon1	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit1	bit 0	Read/ write
LCR[7:0] = 1011 1111	101	Xon2	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit1	bit 0	Read/ write
	110	Xoff1	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit1	bit 0	Read/ write
	111	Xoff2	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit1	bit 0	Read/ write
EFR[4] = 1 and	110	TCR	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit1	bit 0	Read/ write
MCR[6] = 1	111	TLR	bit 7	bit 6	bit 5	bit 4	bit 3	bit 2	bit1	bit 0	Read/ write
MCR[4] = 0 and MCR[2] = 1	111	FIFORdy	RX FIFO D status	RX FIFO C status	RX FIFO B status	RX FIFO A status	TX FIFO D status	TX FIFO C status	TX FIFO B status	TX FIFO A status	Read

⁽¹⁾ Bits represented by shaded cells can only be modified if EFR[4] is enabled, i.e., if enhanced functions are enabled.

Receiver Holding Register (RHR)

The receiver section consists of the receiver holding register (RHR) and the receiver shift register (RSR). The RHR is actually a 64-byte FIFO. The RSR receives serial data from RX terminal. The data is converted to parallel data and moved to the RHR. The receiver section is controlled by the line control register. If the FIFO is disabled, location zero of the FIFO is used to store the characters. If overflow occurs, characters are lost. The RHR also stores the error status bits associated with each character.

⁽²⁾ Refer to the notes under Table 9 for more register access information.

Transmit Holding Register (THR)

The transmitter section consists of the transmit holding register (THR) and the transmit shift register (TSR). The transmit holding register is actually a 64-byte FIFO. The THR receives data and shifts it into the TSR where it is converted to serial data and moved out on the TX terminal. If the FIFO is disabled, location zero of the FIFO is used to store the byte. Characters are lost if overflow occurs.

FIFO Control Register (FCR)

This is a write-only register which is used for enabling the FIFOs, clearing the FIFOs, setting transmitter and receiver trigger levels, and selecting the type of DMA Signaling. Table 11 shows FIFO control register bit settings.

Table 11. FIFO Control Register (FCR) Bit Settings

BIT NO.	BIT SETTINGS
0	0 = Disable the transmit and receive FIFOs 1 = Enable the transmit and receive FIFOs
1	0 = No change 1 = Clears the receive FIFO and resets it's counter logic to zero. Will return to zero after clearing FIFO.
2	0 = No change 1 = Clears the transmit FIFO and resets it's counter logic to zero. Will return to zero after clearing FIFO.
3	0 = DMA Mode 0 1 = DMA Mode 1
5:4 ⁽¹⁾	Sets the trigger level for the TX FIFO: 00 – 8 spaces 01 – 16 spaces 10 – 32 spaces 11 – 56 spaces
7:6	Sets the trigger level for the RX FIFO: 00 – 1 characters 01 – 4 characters 10 – 56 characters 11 – 60 characters

⁽¹⁾ FCR[5-4] can be modified and enabled only when EFR[4] is set. This is because the transmit trigger level is regarded as an enhanced function.

Line Control Register (LCR)

This register controls the data communication format. The word length, number of stop bits, and parity type are selected by writing the appropriate bits to the LCR. Table 12 shows line control register bit settings.

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Table 12. Line Control Register (LCR) Bit Settings

BIT NO.	BIT SETTINGS
1:0	Specifies the word length to be transmitted or received. 00 – 5 bits 01 – 6 bits 10 – 7 bits 11 – 8 bits
2	Specifies the number of stop bits: 0 – 1 stop bits (Word length = 5, 6, 7, 8) 1 – 1.5 stop bits (Word length = 5) 1 – 2 stop bits (Word length = 6, 7, 8) 3
3	0 = No parity 1 = A parity bit is generated during transmission and the receiver checks for received parity.
4	0 = Odd parity is generated (if LCR[3] = 1) 1 = Even parity is generated (if LCR[3] = 1)
5	Selects the forced parity format (if LCR(3) = 1) If LCR[5] = 1 and LCR[4] = 0 the parity bit is forced to 1 in the transmitted and received data. If LCR[5] = 1 and LCR[4] = 1 the parity bit is forced to 0 in the transmitted and received data.
6	Break control bit. 0 = Normal operating condition 1 = Forces the transmitter output to go low to alert the communication terminal.
7	0 = Normal operating condition 1 = Divisor latch enable

Line Status Register (LSR)

Table 13 shows line status register bit settings.

Table 13. Line Status Register (LSR) Bit Settings

BIT NO.	BIT SETTINGS
0	0 = No data in the receive FIFO 1 = At least one character in the RX FIFO
1	0 = No overrun error 1 = Overrun error has occurred.
2	0 = No parity error in data being read from RX FIFO 1 = Parity error in data being read from RX FIFO
3	0 = No framing error in data being read from RX FIFO 1 = Framing error occurred in data being read from RX FIFO (i.e., received data did not have a valid stop bit)
4	0 = No break condition 1 = A break condition occurred and associated byte is 00. (i.e., RX was low for at least one character time frame).
5	0 = Transmit hold register is NOT empty 1 = Transmit hold register is empty. The processor can now load up to 64 bytes of data into the THR if the TX FIFO is enabled.
6	0 = Transmitter hold AND shift registers are not empty. 1 = Transmitter hold AND shift registers are empty.
7	0 = Normal operation 1 = At least one parity error, framing error or break indication are stored in the receiver FIFO. Bit 7 is cleared when no errors are present in the FIFO.

When the LSR is read, LSR[4:2] reflects the error bits [BI, FE, PE] of the character at the top of the RX FIFO (next character to be read). The LSR[4:2] registers do not physically exist, as the data read from the RX FIFO is output directly onto the output data-bus, DI[4:2], when the LSR is read. Therefore, errors in a character are identified by reading the LSR and then reading the RHR.

LSR[7] is set when there is an error anywhere in the RX FIFO and is cleared only when there are no more errors remaining in the FIFO.

NOTE

Reading the LSR does not cause an increment of the RX FIFO read pointer. The RX FIFO read pointer is incremented by reading the RHR.

Modem Control Register (MCR)

The MCR controls the interface with the modem, data set, or peripheral device that is emulating the modem. Table 14 shows modem control register bit settings.

Table 14. Modem Control Register (MCR) Bit Settings⁽¹⁾

BIT NO.	BIT SETTINGS
0	0 = Force DTR output to inactive (high) 1 = Force DTR output to active (low). In loopback controls MSR[5].
1	0 = Force \overline{RTS} output to inactive (high) 1 = Force \overline{RTS} output to active (low). In loopback controls MSR[4]. If Auto-\overline{RTS} is enabled the \overline{RTS} output is controlled by hardware flow control
2	Disables the FIFORdy register Enable the FIFORdy register. In loopback controls MSR[6].
3	0 = Forces the IRQ(A–D) outputs to high-impedance state 1 = Forces the IRQ(A–D) outputs to the active state. In loopback controls MSR[7].
4	0 = Normal operating mode 1 = Enable local loopback mode (internal) In this mode the MCR[3:0] signals are looped back into MSR[3:0] and the TX output is looped back to the RX input internally.
5	0 = Disable Xon Any function 1 = Enable Xon Any function
6	0 = No action 1 = Enable access to the TCR and TLR registers.
7	0 = Divide by one clock input 1 = Divide by four clock input This bit reflects the inverse of the CLKSEL pin value at the trailing edge of the RESET pulse.

⁽¹⁾ MCR[7:5] can be modified only when EFR[4] is set i.e., EFR[4] is a write enable.



Modem Status Register (MSR)

This 8-bit register provides information about the current state of the control lines from the modem, data set, or peripheral device to the processor. It also indicates when a control input from the modem changes state. Table 15 shows modem status register bit settings.

Table 15. Modem Status Register (MSR) Bit Settings (1)

BIT NO.	BIT SETTINGS
0	Indicates that CTS input (or MCR[1] in loopback) has changed state. Cleared on a read.
1	Indicates that DSR input (or MCR[0] in loopback) has changed state. Cleared on a read.
2	Indicates that \overline{RI} input (or MCR[2] in loopback) has changed state from low to high. Cleared on a read.
3	Indicates that $\overline{\text{CD}}$ input (or MCR[3] in loopback) has changed state. Cleared on a read.
4	This bit is equivalent to MCR[1] during local loop-back mode. It is the complement to the CTS input.
5	This bit is equivalent to MCR[0] during local loop-back mode. It is the complement to the DSR input.
6	This bit is equivalent to MCR[2] during local loop-back mode. It is the complement to the RI input.
7	This bit is equivalent to MCR[3] during local loop-back mode. It is the complement to the $\overline{\text{CD}}$ input.

⁽¹⁾ The primary inputs $\overline{\text{RI}}$, $\overline{\text{CD}}$, $\overline{\text{CTS}}$, $\overline{\text{DSR}}$ are all active low but their registered equivalents in the MSR and MCR (in loopback) registers are active high.

Interrupt Enable Register (IER)

The interrupt enable register (IER) enables each of the six types of interrupt, receiver error, RHR interrupt, THR interrupt, Xoff received, or CTS/RTS change of state from low to high. The INT output signal is activated in response to interrupt generation. Table 16 shows interrupt enable register bit settings.

Table 16. Interrupt Enable Register (IER) Bit Settings⁽¹⁾

BIT NO.	BIT SETTINGS
0	0 = Disable the RHR interrupt 1 = Enable the RHR interrupt
1	0 = Disable the THR interrupt 1 = Enable the THR interrupt
2	0 = Disable the receiver line status interrupt 1 = Enable the receiver line status interrupt
3	0 = Disable the modem status register interrupt 1 = Enable the modem status register interrupt
4	0 = Disable sleep mode 1 = Enable sleep mode
5	0 = Disable the Xoff interrupt 1 = Enable the Xoff interrupt
6	0 = Disable the RTS interrupt 1 = Enable the RTS interrupt
7	0 = Disable the CTS interrupt 1 = Enable the CTS interrupt

⁽¹⁾ IER[7:4] can be modified only if EFR[4] is set, i.e., EFR[4] is a write enable. Re-enabling IER[1] causes a new interrupt, if the THR is below the threshold.

Interrupt Identification Register (IIR)

The IIR is a read-only 8-bit register which provides the source of the interrupt in a prioritized manner. Table 17 shows interrupt identification register bit settings.

Table 17. Interrupt Identification Register (IIR) Bit **Settings**

BIT NO.	BIT SETTINGS
0	0 = An interrupt is pending 1 = No interrupt is pending
3:1	3-Bit encoded interrupt. See Table 16.
4	1 = Xoff/Special character has been detected.
5	CTS/RTS low to high change of state
7:6	Mirror the contents of FCR[0]

The interrupt priority list is illustrated in Table 18.

Table 18. Interrupt Priority List

PRIORITY LEVEL	BIT 5	BIT 4	BIT 3	BIT 2	BIT 1	BIT 0	INTERRUPT SOURCE
1	0	0	0	1	1	0	Receiver line status error
2	0	0	1	1	0	0	Receiver timeout interrupt
2	0	0	0	1	0	0	RHR interrupt
3	0	0	0	0	1	0	THR interrupt
4	0	0	1	0	0	0	Modem interrupt
5	0	1	0	0	0	0	Received Xoff signal/special character
6	1	0	0	0	0	0	CTS, RTS change of state from active (low) to inactive (high)

Enhanced Feature Register (EFR)

This 8-bit register enables or disables the enhanced features of the UART. Table 19 shows the enhanced feature register bit settings.

Table 19. Enhanced Feature Register (EFR) Bit Settings

BIT NO.	BIT SETTINGS
3:0	Combinations of software flow control can be selected by programming bit 3-bit 0. See Table 1.
4	Enhanced functions enable bit. 0 = Disables enhanced functions and writing to IER[7:4], FCR[5:4], MCR[7:5]. 1 = Enables the enhanced function IER[7:4], FCR[5:4], and MCR[7:5] can be modified, i.e., this bit is therefore a write enable.
5	0 = Normal operation 1 = Special character detect. Received data is compared with Xoff-2 data. If a match occurs, the received data is transferred to FIFO and IIR[4] is set to 1 to indicate a special character has been detected.
6	RTS flow control enable bit 0 = Normal operation 1 = RTS flow control is enabled i.e., RTS pin goes high when the receiver FIFO HALT trigger level TCR[3:0] is reached, and goes low when the receiver FIFO RESTORE transmission trigger level TCR[7:4] is reached.
7	CTS flow control enable bit 0 = Normal operation 1 = CTS flow control is enabled i.e., transmission is halted when a high signal is detected on the CTS pin.

Product Folder Link(s): TL16C752C



Divisor Latches (DLL, DLH)

Two 8-bit registers store the 16-bit divisor for generation of the baud clock in the baud rate generator. DLH, stores the most significant part of the divisor. DLL stores the least significant part of the division.

DLL and DLH can only be written to before sleep mode is enabled (i.e., before IER[4] is set).

Transmission Control Register (TCR)

This 8-bit register is used to store the receive FIFO threshold levels to start/stop transmission during hardware/software flow control. Table 20 shows transmission control register bit settings.

Table 20. Transmission Control Register (TCR) Bit Settings

BIT NO.	BIT SETTINGS
3:0	RCV FIFO trigger level to HALT transmission (0–60)
7:4	RCV FIFO trigger level to RESTORE transmission (0–60)

TCR trigger levels are available from 0-60 bytes with a granularity of four.

TCR can be written to only when EFR[4] = 1 and MCR[6] = 1. The programmer must program the TCR such that TCR[3:0] > TCR[7:4]. There is no built-in hardware check to make sure this condition is met. Also, the TCR must be programmed with this condition before Auto-RTS or software flow control is enabled to avoid spurious operation of the device.

Trigger Level Register (TLR)

This 8-bit register is used to store the transmit and received FIFO trigger levels used for DMA and interrupt generation. Trigger levels from 4-60 can be programmed with a granularity of 4. Table 21 shows trigger level register bit settings.

Table 21. Trigger Level Register (TLR) Bit Settings

BIT NO.	BIT SETTINGS
3:0	Transmit FIFO trigger levels (4–60), number of spaces available
7:4	RCV FIFO trigger levels (4–60), number of characters available

TLR can be written to only when EFR[4] = 1 and MCR[6] = 1. If TLR[3:0] or TLR[7:4] are zero, then the selectable trigger levels via the FIFO control register (FCR) are used for the transmit and receive FIFO trigger levels. Trigger levels from 4–60 bytes are available with a granularity of four. The TLR should be programmed for N/4, where N is the desired trigger level.

FIFO Ready Register

The FIFO ready register provides real-time status of the transmit and receive FIFOs of both channels. Table 22 shows the FIFO ready register bit settings. The trigger level mentioned in Table 22 refers to the setting in either FCR (when TLR value is zero), or TLR (when it has a nonzero value).

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Table 22. FIFO Ready Register

BIT NO.	BIT SETTINGS
0	0 = There are fewer than a TX trigger level number of spaces available in the TX FIFO of channel A. 1 = There are at least a TX trigger level number of spaces available in the TX FIFO of channel A.
1	0 = There are fewer than a TX trigger level number of spaces available in the TX FIFO of channel B. 1 = There are at least a TX trigger level number of spaces available in the TX FIFO of channel B.
3:2	Unused, always 0
4	0 = There are fewer than a RX trigger level number of characters in the RX FIFO of channel A. 1 = The RX FIFO of channel A has more than a RX trigger level number of characters available for reading or a timeout condition has occurred.
5	0 = There are fewer than a RX trigger level number of characters in the RX FIFO of channel B. 1 = The RX FIFO of channel B has more than a RX trigger level number of characters available for reading or a timeout condition has occurred.
7:6	Unused, always 0

The FIFORdy register is a read only register and can be accessed when any of the two UARTs are selected. CSA or CSB = 0, MCR[2] (FIFORdy Enable) is a logic 1, and loopback is disabled. Its address is 111.

Alternate Function Register (AFR)

The alternate function register (AFR) is used to enable some extra functionality beyond the capabilities of the original TL16C752B. The first of these is a concurrent write mode, which can be useful in more expediently setting up all four UART channels. The second addition is the IrDA mode, which supports Standard IrDA (SIR) mode with baud rates from 2400 to 115.2 bps. The third addition is support for RS-485 bus drivers or transceivers by providing an output pin (DTRx) per channel, which is timed to keep the RS-485 driver enabled as long as transmit data is pending.

The AFR is located at A[2:0] = 010 when LCR[7:5] = 100.

Table 23. Alternate Function Register (AFR) Bit Settings

BIT NO.	BIT SETTINGS
0	CONC enables the concurrent write of all four (754) or two (752) channels simultaneously, which helps speed up initialization. Ensure that any indirect addressing modes have been enabled before using.
1	IREN enables the IrDA SIR mode. This mode is only specified to 115.2 bps and use of this mode at higher speeds is not recommended.
2	485EN enables the half duplex RS-485 mode and causes the DTRx output to be set high whenever there is any data in the THR or TSR and to be held high until the delay set by DLY3:0 has expired, at which time it will be set low. The DTRx output is intended to drive the enabled input of an RS-485 driver. When this bit is set, the transmitter interrupts will be held off until the TSR is empty, unless 485LG is set.
3	485LG is set when the 485EN is set. This bit indicates that a relatively large data block is being set, requiring more than a single load of the xmt fifo. In this case, the transmitter interrupts occur as in the standard RS-232 mode, either when the xmt fifo contents drop below the xmt threshold or when the xmt fifo is empty.
4	RCVEN is valid only when 485EN or IREN is set, and allows the serial receiver to listen in or snoop on the RS485 traffic or IrDA traffic. RS485 mode is generally considered half duplex, and usually a node is either driving or receiving, but there can be cases when it is advantageous to verify what you are sending. This can be used to detect collisions or as part of an arbitration mechanism on the bus. When both RCVEN and 485EN are set, the receiver will store any data presented on RX, if any. Note that implies that the external RS485 receiver is enabled. Whenever 485EN is cleared, the serial receiver is enabled for normal full duplex RS232 traffic. If RCVEN is cleared while 485EN is set, the receiver will be disabled while that channel is transmitting. Standard IrDA (SIR) is also considered half duplex. Often the light energy from the transmitting LED is coupled back into the receiving PIN diode, which creates an input data stream that is not of interest to the host. Disabling the receiver (clearing RCVEN) prevents this reception, and eliminates the task of unloading the data. On the other hand, for diagnostic or other purposes, it may be useful to observe this data stream. For example, a mirror could be used to intentionally couple the output LED to the input PIN. For these cases, RCVEN could be set to enable the receiver. NOTE: When RCVEN is cleared (set to 0), the character timeout interrupt is not available, even in RSA-232 mode. This can be useful when checking code for valid threshold interrupts, as the timeout interrupt will not override the threshold interrupt.
7:5	DLY3–DLY0 sets a delay after the last stop bit of the last data byte being set before the $\overline{\text{DTR}}$ x is set low, to allow for long cable runs. The delay is in number of bit times and is enabled by 485EN. The delay will start only when both the xmt serial shift register (TSR) is empty and the xmt fifo (THR) is empty, and if started, will be cleared by any data being written to the THR.

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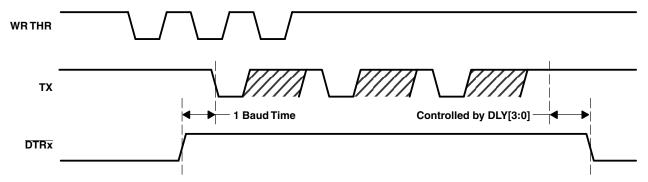
Table 24. LOOP and RCVEN Functionality

LOOP MODE	RCVEN	AFR	MODE	DESCRIPTION				
	RCVEN = 1	AFR = 10	RS-232	Receive threshold, timeout, and error detection interrupts available. Data stored in receive FIFO.				
		AFR = 14	RS-485	Receive threshold, timeout, and error detection interrupts available. Data stored in receive FIFO.				
LOOP mode off, MCR4 = 0, RX, TX active		AFR = 12	IrDA	Receive threshold, timeout, and error detection interrupts available. Data stored in receive FIFO.				
KA, IA active	RCVEN = 0	AFR = 00	RS-232	Receive threshold and error detection interrupts available. Data stored in receive FIFO.				
		AFR = 04	FR = 04 RS-485 No data stored in receive FIFO, hence no interrupts a					
		AFR = 02	IrDA	No data stored in receive FIFO, hence no interrupts available.				
	RCVEN = 1			RS-232	Receive threshold, timeout, and error detection interrupts available. Data stored in receive FIFO.			
		AFR = 14	RS-485	Receive threshold, timeout, and error detection interrupts available. Data stored in receive FIFO.				
LOOP mode on,		AFR = 12	IrDA	Receive threshold, timeout, and error detection interrupts available. Data stored in receive FIFO.				
RX, TX inactive	MCR4 = 1, , TX inactive		RS-232	Receive threshold and error detection interrupts available. Data stored in receive FIFO.				
	RCVEN = 0	AFR = 04	RS-485	Receive threshold and error detection interrupts available. Data stored in receive FIFO.				
		AFR = 02	IrDA	Receive threshold and error detection interrupts available. Data stored in receive FIFO.				

RS-485 Mode

The RS-485 mode is intended to simplify the interface between the UART channel and an RS-485 driver or transceiver. When enabled by setting 485EN, the DTRx output goes high one bit time before the first stop bit of the first data byte being sent, and remains high as long as there is pending data in the transmitter shift register (TSR) or transmitter holding register (THR, xmt fifo). Once both are empty (after the last stop bit of the last data byte), the DTRx output stays high for a programmable delay of 0 to 15 bit times, as set by DLY[3:0]. This helps preserve data integrity over long signal lines. This is illustrated in the following.

Often RS-485 packets are relatively short and the entire packet can fit within the 64 byte xmt fifo. In this case, it goes empty when the TSR goes empty. But in cases where a larger block needs to be sent, it is advantageous to reload the xmt fifo as soon as it is depleted. Otherwise, the transmission stalls while waiting for the xmt fifo to be reloaded, which varies with processor load. In this case, it is best to also set 485LG (large block), which causes the transmit interrupt to occur wither when the THR becomes empty (if the xmt fifo level was not above the threshold), or when the xmt fifo threshold is crossed. The reloading of the xmt fifo occurs while some data is being shifted out, eliminating fifo underrun. If desired, when the last bytes of a current transmission are being loaded in the xmt fifo, 485LG can be cleared before the load and the transmit interrupt occurs on the TSR going empty.



A. Waveforms are not shown to scale, as the WR THR pulses typically are less than 100 ns, where the TX waveform varies with baud rate but is typically in the microsecond range.

Figure 24. DTRx and Transmit Data Relationship

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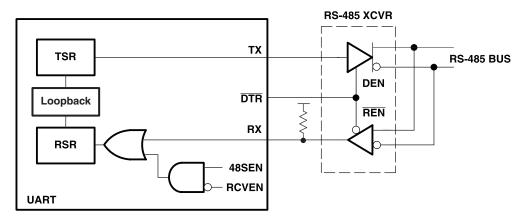


Figure 25. RS-485 Application Example 1

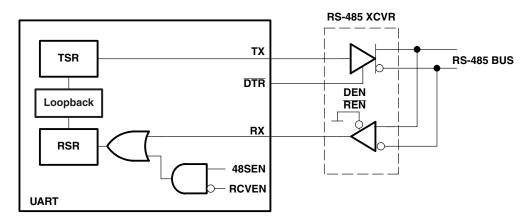


Figure 26. RS-485 Application Example 2

IrDA Overview

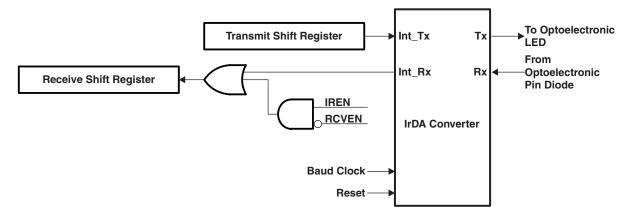


Figure 27. IrDA Mode

The infrared data association (IrDA) defines several protocols for sending and receiving serial infrared data, including rates of 115.2 kbps, 0.576 Mbps, 1.152 Mbps, and 4 Mbps. The low rate of 115.2 kbps was specified first and the others must maintain downward compatibility with it. At the 115.2 kbps rate, the protocol implemented in the hardware is fairly simple. It primarily defines a serial infrared data word to be surrounded by a start bit equal to 0 and a stop bit equal to 1. Individual bits are encoded or decoded the same whether they are

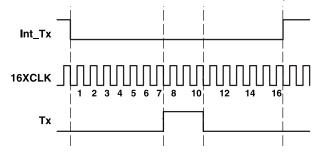
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start, data, or stop bits. The IrDA engine in the TL16C752C evaluate only single bits and only follow the 115.2 kbps protocol. The 115.2 kbps rate is a maximum rate. When both ends of the transfer are set up to a lower but matching speed, the protocol still works. The clock used to code or sample the data is 16 times the baud rate, or 1.843 MHz maximum. To code a 1, no pulse is sent or received for 1-bit time period, or 16 clock cycles. To code a 0, one pulse is sent or received within a 1-bit time period, or 16 clock cycles. The pulse must be at least 1.6 µs wide and 3 clock cycles long at 1.843 MHz. At lower baud rates the pulse can be 1.6 µs wide or as long as 3 clock cycles. The transmitter output, Tx, is intended to drive a LED circuit to generate an infrared pulse. The LED circuits work on positive pulses. A terminal circuit is expected to create the receiver input, Rx. Most, but not all, PIN circuits have inversion and generate negative pulses from the detected infrared light. Their output is normally high. The TL16C752C can decode either negative or positive pulses on Rx.

IrDA Encoder Function

Serial data from a UART is encoded to transmit data to the optoelectronics. While the serial data input to this block (Int_Tx) is high, the output (Tx) is always low, and the counter used to form a pulse on Tx is continuously cleared. After Int_Tx resets to 0, Tx rises on the falling edge of the seventh 16XCLK. On the falling edge of the tenth 16XCLK pulse, Tx falls, creating a 3-clock-wide pulse. While Int_Tx stays low, a pulse is transmitted during the seventh to tenth clocks of each 16-clock bit cycle.



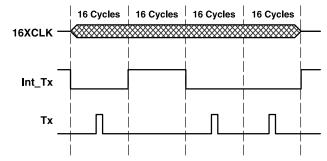
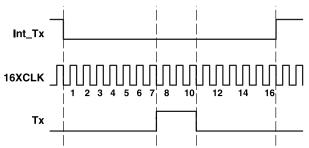


Figure 28. IrDA-SIR Encoding Scheme – Detailed Timing Diagram

Figure 29. Encoding Scheme - Macro View

After reset, Int_Rx is high and the 4-bit counter is cleared. When a falling edge is detected on Rx, Int_Rx falls on the next rising edge of 16XCLK with sufficient setup time. Int_Rx stays low for 16 cycles (16XCLK) and then returns to high as required by the IrDA specification. As long as no pulses (falling edges) are detected on Rx, Int_Rx remains high.



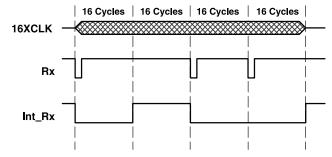


Figure 30. IrDA-SIR Decoding Scheme – Detailed Timing Diagram

Figure 31. IrDA-SIR Decoding Scheme – Macro View

It is possible for jitter or slight frequency differences to cause the next falling edge on Rx to be missed for one 16XCLK cycle. In that case, a 1-clock-wide pulse appears on Int_Rx between consecutive zeroes. It is important for the UART to strobe Int_Rx in the middle of the bit time to avoid latching this 1-clock-wide pulse. The TL16C550C UART already strobes incoming serial data at the proper time. Otherwise, note that data is required to be framed by a leading zero and a trailing one. The falling edge of that first zero on Int_Rx synchronizes the read strobe. The strobe occurs on the eighth 16XCLK pulse after the Int_Rx falling edge and once every 16 cycles thereafter until the stop bit occurs.

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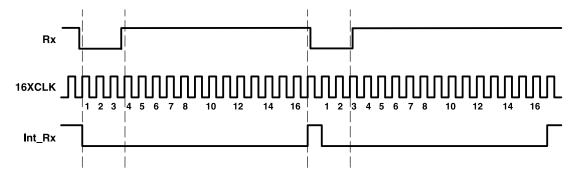


Figure 32. Timing Causing 1-Clock-Wide Pulse Between Consecutive Ones

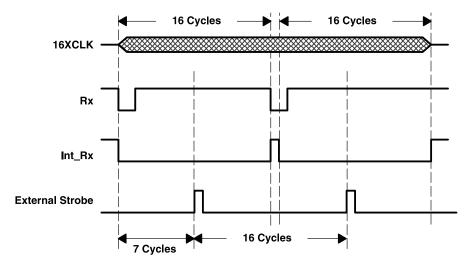


Figure 33. Recommended Strobing for Decoded Data

The TL16C752C can decode positive pulses on Rx. The timing is different, but the variation is invisible to the UART. The decoder, which works from the falling edge, now recognizes a zero on the trailing edge of the pulse rather than on the leading edge. As long as the pulse width is fairly constant, as defined by the specification, the trailing edges should also be 16 clock cycles apart and data can readily be decoded. The zero appears on Int_Rx after the pulse rather than at the start of it.

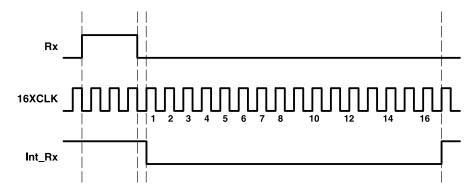


Figure 34. Positive Rx Pulse Decode – Detailed View



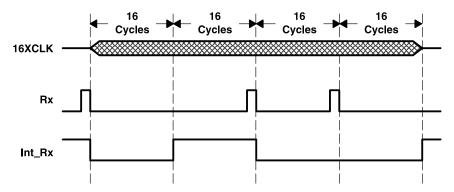


Figure 35. Positive Rx Pulse Decode – Macro View

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TL16C752C Programmer's Guide

The base set of registers that are used during high-speed data transfer have a straightforward access method. The extended function registers require special access bits to be decoded along with the address lines. The following guide will help with programming these registers. Note that the descriptions below are for individual register access. Some streamlining through interleaving can be obtained when programming all the registers.

Set baud rate to VALUE1, VALUE2 Read LCR (03), save in temp

Set LCR (03) to 80 Set DLL (00) to VALUE1 Set DLM (01) to VALUE2 Set LCR (03) to temp

Set Xoff1,Xon1 to VALUE1,VALUE2 Read LCR (03), save in temp

Set LCR (03) to BF Set Xoff1 (06) to VALUE1 Set Xon1 (04) to VALUE2 Set LCR (03) to temp

Set Xoff2,Xon2 to VALUE1,VALUE2 Read LCR (03), save in temp

Set LCR (03) to BF Set Xoff2 (07) to VALUE1 Set Xon2 (05) to VALUE2 Set LCR (03) to temp

Read LCR (03), save in temp Set software flow control mode to VALUE

Set LCR (03) to BF Set EFR (02) to VALUE Set LCR (03) to temp

Set flow control threshold to VALUE Read LCR (03), save in temp1

Set LCR (03) to BF

Read EFR (02), save in temp2 Set EFR (02) to 10 + temp2

Set LCR (03) to 00

Read MCR (04), save in temp3 Set MCR (04) to 40 + temp3 Set TCR (06) to VALUE Set LCR (03) to BF Set EFR (02) to temp2 Set LCR (03) to temp1 Set MCR (04) to temp3

Set xmt and rcv FIFO thresholds to VALUE Read LCR (03), save in temp1

Set LCR (03) to BF

Read EFR (02), save in temp2 Set EFR (02) to 10 + temp2 Set LCR (03) to 00 Read MCR (04), save in temp3 Set MCR (04) to 40 + temp3 Set TLR (07) to VALUE Set LCR (03) to BF Set EFR (02) to temp2 Set LCR (03) to temp1 Set MCR (04) to temp3

Read FIFORdy register Read MCR (04), save in temp1

Set temp2 = temp1 * EF Set MCR (04), save in temp2 Read FRR (07), save in temp2 Pass temp2 back to host Set MCR (04) to temp1

PACKAGE OPTION ADDENDUM

www.ti.com 28-May-2009

PACKAGING INFORMATION

Orderable Device	Status ⁽¹⁾	Package Type	Package Drawing	Pins	Package Qty	e Eco Plan ⁽²⁾	Lead/Ball Finish	MSL Peak Temp ⁽³⁾
TL16C752CIPFB	ACTIVE	TQFP	PFB	48	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TL16C752CIPFBR	ACTIVE	TQFP	PFB	48	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TL16C752CIRHBR	ACTIVE	QFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
TL16C752CIRHBRG4	ACTIVE	QFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
TL16C752CPFB	ACTIVE	TQFP	PFB	48	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TL16C752CPFBR	ACTIVE	TQFP	PFB	48	1000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR
TL16C752CRHBR	ACTIVE	QFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR
TL16C752CRHBRG4	ACTIVE	QFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-3-260C-168 HR

⁽¹⁾ The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

(3) MSL, Peak Temp. -- The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

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TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TL16C752CIPFBR	TQFP	PFB	48	1000	330.0	16.4	9.6	9.6	1.5	12.0	16.0	Q2
TL16C752CIRHBR	QFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.5	8.0	12.0	Q2
TL16C752CPFBR	TQFP	PFB	48	1000	330.0	16.4	9.6	9.6	1.5	12.0	16.0	Q2
TL16C752CRHBR	QFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.5	8.0	12.0	Q2





*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TL16C752CIPFBR	TQFP	PFB	48	1000	346.0	346.0	33.0
TL16C752CIRHBR	QFN	RHB	32	3000	340.5	333.0	20.6
TL16C752CPFBR	TQFP	PFB	48	1000	346.0	346.0	33.0
TL16C752CRHBR	QFN	RHB	32	3000	340.5	333.0	20.6

RHB (S-PVQFN-N32)

PLASTIC QUAD FLATPACK NO-LEAD



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.

- B. This drawing is subject to change without notice.
- C. QFN (Quad Flatpack No-Lead) Package configuration.
- D. The package thermal pad must be soldered to the board for thermal and mechanical performance.
- E. See the additional figure in the Product Data Sheet for details regarding the exposed thermal pad features and dimensions.
- F. Falls within JEDEC MO-220.



RHB (S-PVQFN-N32)

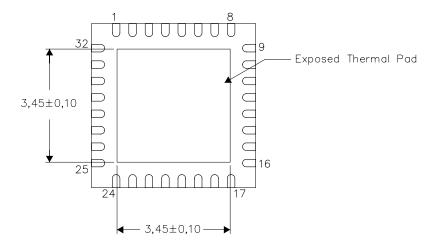
PLASTIC QUAD FLATPACK NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

The exposed thermal pad dimensions for this package are shown in the following illustration.



Bottom View

Exposed Thermal Pad Dimensions

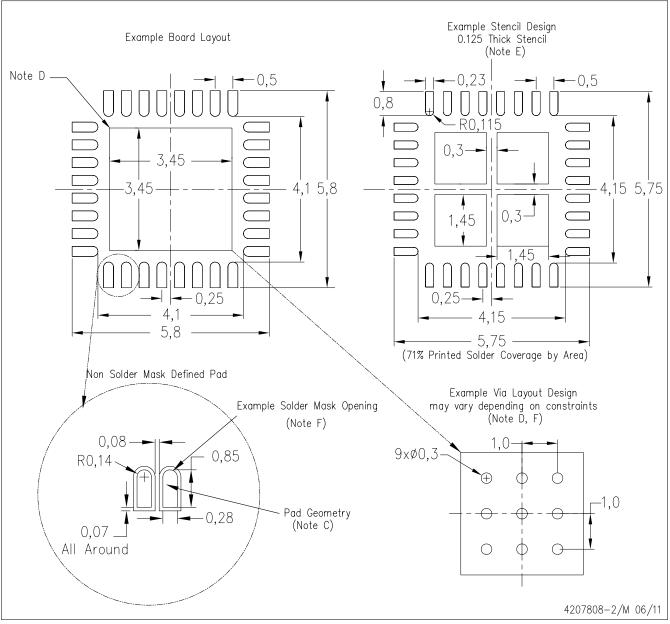
4206356-2/U 06/11

NOTE: A. All linear dimensions are in millimeters



RHB (S-PVQFN-N32)

PLASTIC QUAD FLATPACK NO-LEAD



NOTES:

- A. All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, Quad Flat—Pack Packages, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com http://www.ti.com>.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for recommended solder mask tolerances and via tenting recommendations for vias placed in the thermal pad.



PFB (S-PQFP-G48)

PLASTIC QUAD FLATPACK



NOTES: A. All linear dimensions are in millimeters.

B. This drawing is subject to change without notice.

C. Falls within JEDEC MS-026

PFB (S-PQFP-G48)



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



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